

AGH University of Science and Technology

Faculty of Electrical Engineering, Automatics, Computer Science and Biomedical Engineering

Department of Measurement and Electronics

Ph. D. dissertation

# DIGITALLY-ASSISTED ANALOG CIRCUITS FOR HYBRID PIXEL X-RAY DETECTORS

mgr inż. Aleksandra Krzyżanowska

Ph. D. advisor: dr hab. inż. Grzegorz Deptuch

Kraków, 2018



Akademia Górniczo-Hutnicza im. Stanisława Staszica w Krakowie Wydział Elektrotechniki, Automatyki, Informatyki i Inżynierii Biomedycznej Katedra Metrologii i Elektroniki

Rozprawa doktorska

# WSPOMAGANE CYFROWO UKŁADY ANALOGOWE DLA POTRZEB HYBRYDOWYCH DETEKTORÓW PIKSELOWYCH PROMIENIOWANIA X

mgr inż. Aleksandra Krzyżanowska

Promotor: dr hab. inż. Grzegorz Deptuch

Kraków, 2018

#### Abstract

The dissertation describes the theoretical studies and experimental tests of the multichannel readout circuits for hybrid X-ray detectors, addressing aspects of the design of integrated circuits which contain both analog circuitry as well as digital logic integrated on the same silicon substrate. Combining more and more functionality inside a pixel is a trend observed in nowadays readout integrated circuits. However, integration of several blocks of diverse functionality in each readout channel and implementation of inter-pixel communication impose new challenges in the process of simulation, design, implementation and verification of the integrated mixed-mode systems.

Scientific problems addressed by this dissertation aim at improving design process of mixedmode readout integrated circuits, in particular, with the algorithms dealing with charge sharing implemented inside a chip. To achieve this goal, the basic concepts of X-ray detector systems and their limitations are studied. Then, the algorithms implemented in mixed-mode integrated circuits for dealing with charge sharing and selected solutions from the literature are discussed. The C8P1 algorithm, developed by the ASIC group from the AGH University of Science and Technology, is a subject of detailed conceptual analyses and simulations as a known solution to the charge sharing problem. The simulation approach to this algorithm and implementation of realistic models, including practical aspects, for example non-ideal comparators, noise or analog parameters spread is presented. The influence of analog parameters spread on the detector registration is analysed. The simulations are conducted in static and dynamic modes.

The target of the experimental part of the thesis is to confirm the conclusions obtained through the simulations. Firstly, the architecture and design aspects of the multichannel readout integrated circuit named Chase Jr. chip with the C8P1 algorithm are presented. The operation and configuration of the chip with an emphasis on analog path reconfiguration for the purpose of testing and trimming is revealed. The dedicated measurement environment implemented for the Chase Jr. tests and the practical realisation of the testing procedures for the Chase Jr. chip are described. The measurement results for the Chase Jr. chip bonded to a silicon sensor are shown. Three types of experiments are conducted: the preliminary integrated circuit tests without X-ray radiation performed for calibration purposes, the experiments using an X-ray tube and the experiments using specialised synchrotron source, used mainly for the assessment of the algorithm dealing with charge sharing in this work. The experiments include the tests of signal reconstruction in the case of charge sharing, the tests of registration at pixel borders, the tests of influence of correction on the C8P1 algorithm and the high count rate tests.

The results of the simulations and measurements lead to the conclusions that the integrated circuit with the C8P1 algorithm switched on allows reconstruction of the total photon energy from fractional signals in the case of charge sharing between two or four pixels, and thus, the photons

can be detected even at pixel borders, where the standard approach fails. This proves that the charge sharing effect occurring in hybrid pixel detectors can be compensated by mixed analog-digital circuits implemented inside the readout electronics using inter-pixel communication strategies. However, increasing pixel-to-pixel gain spread, DC offset spread and noise, result in the significant degradation of the detection efficiency in the C8P1 mode. Therefore, there is a need for dedicated correction circuits to minimise the analog parameters spread between channels and assure the proper operation of the detector. The results of the tests of the Chase Jr. chip prove in practice, that it is possible to overcome technology limitations regarding analog parameters spread of the multichannel integrated circuits for hybrid pixel detectors with inter-pixel communication, using digitally assisted correction blocks.

#### Abstrakt

Niniejsza rozprawa przedstawia rozważania teoretyczne, wyniki symulacji oraz pomiarów wielokanałowych układów odczytowych dla hybrydowych detektorów promieniowania X. Układy takie zawierają zarówno część analogową jak i cyfrową, zintegrowane w jednym układzie scalonym. Widocznym trendem w projektowaniu tych układów jest zwiększanie funkcjonalności pojedynczego kanału, co stawia nowe wyzwania w procesie symulacji, projektowania i testowania zintegrowanych systemów analogowo-cyfrowych.

Podjęte w ramach tej pracy rozważania naukowe mają na celu optymalizację procesu projektowania analogowo-cyfrowych scalonych układów odczytowych, w szczególności zawierających algorytmy do minimalizacji efektów związanych ze zjawiskiem podziału ładunku. We wstępie wprowadzono podstawowe pojęcia, opisano zjawiska i rozwiązania dedykowane detekcji promieniowania X. Następnie, przedstawiono algorytmy i implementacje układowe podejmujące próby rozwiązania problemu podziału ładunku w detektorach pracujących w trybie pojedynczego zliczania fotonów opisane w literaturze. Przedmiotem szerszych badań koncepcyjnych i symulacyjnych stał się algorytm C8P1, zaproponowany przez grupę projektowania układów scalonych z Katedry Metrologii i Elektroniki, Akademii Górniczo-Hutniczej. W pracy zaproponowano statyczne i dynamiczne modele układu odczytowego uwzględniające m.in. rozrzuty parametrów analogowych pomiędzy kanałami, wynikające z efektów niedopasowania technologicznego oraz szumy.

W eksperymentalnej części pracy zaprezentowano architekturę, zasadę działania oraz konfigurację układu scalonego Chase Jr. z zaimplementowanym algorytmem C8P1. W celu przeprowadzenia zautomatyzowanych testów samego układu oraz detektora składającego się z układu Chase Jr. zbondowanego do krzemowego czujnika, zaprojektowano oraz zaimplementowano dedykowane środowisko testowe. Przeprowadzono trzy rodzaje eksperymentów: wstępne testy układu scalonego bez promieniowania X w celach kalibracyjnych, testy z użyciem lampy rentgenowskiej oraz testy synchrotronowe. Przedmiotem szerszej analizy były procedury korekcyjne i wpływ korekcji na poprawność detekcji, rekonstrukcja sygnału w detektorze w przypadku podziału ładunku oraz działanie detektora w warunkach promieniowania o dużym natężeniu.

Zarówno wyniki symulacji jak i pomiarów pokazują, że układ scalony z zaimplementowanym algorytmem C8P1 pozwala na odtworzenie pierwotnej energii padającego fotonu w przypadku podziału ładunku pomiędzy dwa lub cztery piksele, zatem fotony mogą zostać zarejestrowane nawet na krawędziach pomiędzy pikselami. Oznacza to, że podział ładunku może zostać skompensowany przy wykorzystaniu układów analogowych wspomaganych cyfrowo wykorzystujących komunikację międzypikselową. Jednakże, zaobserwowano i udowodniono, że znaczący wpływ na jakość detekcji mają szumy elektroniki odczytu oraz rozrzut parametrów analogowych układu, takich jak offset DC na wejściu dyskryminatora czy wzmocnienie. Wskazano na konieczność stosowania dedykowanych układów i procedur korekcyjnych do minimalizacji rozrzutów pomiędzy kanałami, co jest warunkiem poprawnej pracy detektora. Wyniki uzyskane w testach układu Chase Jr. bezpośrednio udowadniają, że możliwe jest pokonanie ograniczeń technologicznych dotyczących rozrzutów parametrów analogowych w hybrydowych pikselowych detektorach promieniowania X, wykorzystując cyfrowe wspomaganie bloków analogowych.

#### **ACKNOWLEDGEMENTS**

This work was supported by the Polish National Science Center under Contract UMO-2013/09/B/ST7/01627 and tests of inter-pixel communication was supported by Contract DEC-2014/13/B/ST7/01168.

Aleksandra Krzyżanowska would like to thank Polish Ministry of Science and Higher Education for financial support.

The synchrotron X-ray measurements were carried out at the beamline 1-BM of the Advanced Photon Source at the Argonne National Laboratories operated by the U.S. Department of Energy, Office of Basic Energy Sciences. The author would like to thank A. Macrander and S. Stoupin for their help with the beamline environment. The X-ray tube measurements were carried out at the Rigaku Co., Tokio, Japan.

I would like to express my gratitude to many people who have helped me during my research. First and foremost I want to thank my supervisor, Grzegorz Deptuch, for his motivation, support and patience. I am also grateful to my co-advisor, Piotr Maj who was always willing to share his knowledge and experience. I would like to thank Paweł Gryboś, Robert Szczygieł, Krzysztof Kasiński, Rafał Kłeczek, Piotr Otfinowski, Piotr Kmon, Mirosław Żołądź, Anna Kozioł, Jacek Rauza for their help and support. I would also like to thank Krzysztof Krawczyk, Andrzej Wajda, Teresa Michniak for their work and all the members of the ASIC group and colleagues from the Department of Measurement and Electronics for their positive attitude and feedback. Last but not least, I wish to thank my family and friends for their support.

Dziękuję moim najbliższym za wsparcie i pomoc. Profesor Vetulani mawiał, że kariera naukowa to chęć bycia uznanym przez garstkę ludzi na świecie, o których nikt poza nami nie słyszał. Zatem, dziękuję Wam, że wierzycie mi na słowo, że moja praca ma sens.

## TABLE OF CONTENTS

1 INTRODUCTION	22
1.1 WORK THESIS	22
1.2 X-RAY GENERATION	25
1.2.1 X-RAY TUBE	25
1.2.2 Synchrotron radiation source	26
1.2.3 X-RAY SOURCES IN TESTS OF 2D POSITION SENSITIVE DEVICES	28
1.3 X-RAY INTERACTION WITH MATTER	30
1.3.1 DOMINANT MECHANISMS OF INTERACTION OF X-RAY PHOTONS WITH ATOMS	30
1.3.2 ELECTRON - HOLE PAIRS GENERATION IN SEMICONDUCTOR SENSOR MATERIALS	33
1.3.3 BEAM ATTENUATION AND FILTRATION	34
1.4 X-RAY IMAGING IN APPLICATIONS	34
1.4.1 X-RAY RADIOGRAPHY	35
1.4.2 X-RAY SPECTROSCOPY	35
1.4.3 X-RAY DIFFRACTION AND SCATTERING	
1.4.4 TIME-RESOLVED X-RAY EXPERIMENTS	
<b>1.5 OPERATION OF HYBRID SEMICONDUCTOR POSITION SENSITIVE DETECTOR FOR X-RAY-IMAGING</b>	37
1.5.1 CHARGE TRANSPORT IN SENSOR - DRIFT AND DIFFUSION	
1.5.2 Charge sharing effect	
1.5.3 OTHER EFFECTS DEGRADING PERFORMANCE OF PHOTON COUNTING DETECTORS	41
1.5.4 Common sensor materials	41
1.5.5 READOUT INTEGRATED CIRCUIT	
1.5.6 METHODS OF ESTIMATION OF ANALOG CHAIN PARAMETERS	43
1.6 LIMITATIONS IN READOUT ELECTRONICS	45
1.6.1 МІЅМАТСН	46
1.6.2 Electronic noise	49
1.6.3 PULSE PILE-UP	51
1.6.4 Crosstalk	53
1.6.5 CONCLUSIONS ON INTEGRATED CIRCUIT LIMITATIONS	54

αρατάτνας και μτιαν	55

2.1 ALGORITHM CONCEPTS FOR DEALING WITH CHARGE SHARING	55
2.2 SELECTED ASICS	56
2.2.1 Pixirad Pixie III	57
2.2.2 X-Counter	59
2.2.3 MEDIPIX 3/3RX	59
2.3 PROTOTYPE SOLUTION - C8P1 IN MINIVIPIC AND CHASE JR. CHIP	62
2.3.1 C8P1 ALGORITHM CONCEPT	62
2.3.2 Two implementations of C8P1 algorithm	64
2.3.3 EXAMPLE CASES OF SIGNAL PROCESSING IN C8P1 ALGORITHM	67
2.4 COMPARISON OF C8P1 AND OTHER EXISTING SOLUTIONS	71

#### 

3.1 ANALYTICAL APPROACH TOWARDS MODELLING OF READOUT CHANNEL	
3.2 STATIC MODEL OF DETECTOR	
3.2.1 MODEL IMPLEMENTATION FOR MONTE CARLO SIMULATIONS	75
3.2.2 SIMULATION RESULTS	76
3.3 DYNAMIC MODEL OF READOUT CHANNEL	
3.3.1 MODEL IMPLEMENTATION	
3.3.2 SIMULATION RESULTS	
3.4 CONCLUSIONS ON STATIC AND DYNAMIC MODELLING	
4 INTEGRATED CIRCUIT DEALING WITH CHARGE SHARING	
4.1 MAIN ASPECTS OF CHASE JR. CHIP AND DETECTOR MODULE DESIGN	
4.2 TESTABILITY FEATURES OF CHASE JR. CHIP	
4.2.1 INDIVIDUAL PIXELS CONTRIBUTION INTO SUMMING	94
4.2.2 Standard and C8P1 mode of operation	

4.4.1 DC OFFSETS TRIMMING	96
4.4.2 GAINS TRIMMING	

<b>5 DEDICATED MEASUREMENT ENVIRONMENT FOR INTEGRATED CIRCUIT FO</b>	<u>R HYBRID PIXEL</u>
DETECTOR	<u></u>
5.1 REQUIREMENTS FOR MEASUREMENT ENVIRONMENT	
5.2 TEST PROCEDURES	
5.2.1 PRELIMINARY TESTS WITHOUT X-RAY RADIATION	
5.2.2 Tests with X-ray radiation	
5.3 Measurement set-up	
5.4 SOFTWARE FUNCTIONALITY FOR TESTING PROCEDURES	
5.5 SOFTWARE ARCHITECTURE AND IMPLEMENTATION	
6 MEASUREMENTS RESULTS	
6.1 CORRECTION RESULTS	
6.1.1 DC OFFSET CORRECTION	
6.1.2 GAIN CORRECTION	
6.2 EXPERIMENTAL RESULTS OF TESTS WITHOUT X-RAY RADIATION	
6.2.1 LINEARITY OF CALIBRATION CIRCUIT	
6.2.2 Tests of pixels contribution into summing in C8P1 Algorithm	
6.2.3 INFLUENCE OF CORRECTION ON C8P1 RESULTS	
6.2.4 Emulation of pencil beam test using calibration circuit	
6.3 EXPERIMENTAL RESULTS OF TESTS WITH X-RAY RADIATION	
6.3.1 Experiment set-ups	
6.3.2 TESTS OF SIGNAL RECONSTRUCTION IN THE CASE OF CHARGE SHARING	
6.3.3 TESTS OF REGISTRATION AT PIXEL BORDERS	
6.3.4 INFLUENCE OF CORRECTION ON C8P1PERFORMANCE	
6.3.5 High count rate tests	
6.4 CONCLUSIONS ON MEASUREMENT METHODS AND C8P1 ALGORITHM PERFORMANCE	

<b>7 CONCLUSIONS AND PROSPECTS</b>	 6
	_

7.1 CONCLUSIONS FROM SIMULATION RESULTS	138
7.2 CONCLUSIONS FROM EXPERIMENTAL RESULTS	140
7.3 PROSPECTS	141
8 REFERENCES	144

### LIST OF FIGURES

FIG. 1.1 A TYPICAL X-RAY SPECTRUM FOR AN X-RAY TUBE26
FIG. 1.2 A CUTAWAY OF THE APS EXPERIMENT HALL, WITH STORAGE RING COMPONENTS, INSERTION DEVICES, BEAMLINES, FIRST-OPTICS
ENCLOSURES, RESEARCH STATIONS, A TYPICAL LAB/OFFICE MODULE (FIGURE FROM [7])27
FIG. 1.3 A) A CONCEPT OF A FLAT FIELD ILLUMINATION TEST. B) A CONCEPT OF A PENCIL BEAM TEST
Fig. 1.4 The mechanisms of an X-ray photon interaction with matter
FIG. 1.5 THE PHOTON CROSS-SECTION A) FOR SI AND B) FOR CDTE
FIG. 1.6 A SEMICONDUCTOR HYBRID PIXEL DETECTOR
FIG. 1.7 THE OPERATION OF A SEMICONDUCTOR HYBRID PIXEL DETECTOR IN CASE OF CHARGE SHARING
FIG. 1.8 THE PERCENTAGE OF EVENTS AFFECTED BY THE CHARGE SHARING EFFECT AS A FUNCTION OF THE DETECTOR CHANNEL SIZE FOR
A SILICON SENSOR OF THE THICKNESS D=300 $\mu$ M, AND FOR THE BIAS VOLTAGE APPLIED V=100 V40
FIG. 1.9 A STANDARD ARCHITECTURE OF A READOUT CHANNEL OF A HYBRID PIXEL DETECTOR WITH THE ALTERNATIVE METHODS OF
DIGITIZATION
Fig. $1.10~A$ sample threshold scan plot with the threshold voltages $V_{\text{Noise}},V_0$ marked
Fig. 1.11 The technology dependent constants $A_{VT}$ represented with black squares, and $A_B$ represented with diamonds
FOR A) NMOS AND B) PMOS DEVICES (FIGURES FROM [49])47
Fig. 1.12 The standard architecture of a readout channel for a hybrid pixel detector with added DC offset and gain
TRIMMING FEATURES
Fig. 1.13 The ENC components versus shaper peaking time $\tau_{\text{s}}$
Fig. 1.14 The Poisson distributed pulse trains for t = 100 ns, T = 1 $\mu$ s and A) $\lambda$ = 1, b) $\lambda$ = 2, c) $\lambda$ = 10
FIG. 1.15 THE OUTPUT COUNT RATES AS A FUNCTION OF INPUT COUNT RATES FOR A) POISSON DISTRIBUTED PULSES, B) SYNCHRONISED
PULSES
FIG. 2.1 THE PROPOSED HEXAGONAL-PIXEL-DETECTOR LAYOUT (LEFT) AND A PIXEL ARCHITECTURE (RIGHT) REVEALED AS AN
ANNOUNCEMENT OF FUTURE DEVELOPMENTS FOR MEDIPIX2, ACCORDING TO [72].
FIG. 2.2 A SCHEMATIC DIAGRAM OF THE READOUT CHANNEL ARCHITECTURE OF MEDIPIX3 ACCORDING TO [73]60
FIG. 2.3 AN ILLUSTRATION OF THE CHARGE SUMMING ALGORITHM ACCORDING TO [75]. A) THE LOCAL SIGNAL IS COMPARED WITH A
THRESHOLD. THE ARBITRATION NETWORK DETERMINES WHICH PIXEL RECEIVES THE LARGEST CHARGE. B) THE RECONSTRUCTED
SUMMED SIGNAL FROM FOUR PIXELS IS COMPARED WITH A THRESHOLD60
Fig. 2.4 A schematic diagram of the readout channel architecture of Medipix3RX according to [75]61
FIG. 2.5 THE C8P1 ALGORITHM: RECONSTRUCTION OF THE TOTAL CHARGE. FIGURE FROM [67]62

FIG. 2.6 A SCHEMATIC DIAGRAM OF THE READOUT CHANNEL ARCHITECTURE WITH THE C8P1 ALGORITHM.	63
FIG. 2.7 THE LOGIC SIGNALS USED BY THE C8P1 BLOCK TO DETERMINE IF THE HIT COUNTER IN THE PIXEL SHOULD BE INCREM	MENTED64
FIG. 2.8 A SCHEMATIC DIAGRAM OF THE READOUT CHANNEL ARCHITECTURE OF THE MINIVIPIC CHIP	65
FIG. 2.9 A SCHEMATIC DIAGRAM OF THE READOUT CHANNEL ARCHITECTURE OF THE CHASE JR. CHIP.	
FIG. 2.10 A) A TIMING DIAGRAM OF A SHAPER FAST SIGNAL AND SAMPLE THRESHOLD SETTINGS. B) A TIMING DIAGRAM OF TH	E RESULTING
'ACTIVATE PIX' AND 'HIT' SIGNALS FOR GIVEN THRESHOLDS	
FIG. 2.11 A) A TIMING DIAGRAM OF THE SIGNALS USED TO DETERMINE THE C8P1 ALGORITHM RESULT. B) A SCHEMATIC DIAG	GRAM OF THE
DESIGN OF THE 'C8P1 AND PULSE STRETCHING BLOCK'	
FIG. 2.12 THE C8P1 CONCEPT WHEN A PHOTON HITS THE DETECTOR IN THE PIXEL CENTRE. 100% ELECTRONS COLLECTED F	3Y THE PIXEL
P22	
FIG. 2.13 THE C8P1 CONCEPT WHEN A PHOTON HITS THE DETECTOR IN BETWEEN TWO PIXELS, AND 70% AND 30% OF ELE	CTRONS ARE
COLLECTED BY THE PIXEL P22 AND P32, RESPECTIVELY	
FIG. 2.14 THE C8P1 CONCEPT WHEN A PHOTON HITS THE DETECTOR IN BETWEEN FOUR PIXELS. 70%, 13%, 12%	AND 5% OF
ELECTRONS COLLECTED BY THE PIXEL P22, P32, P23 AND P33 RESPECTIVELY.	70
FIG. 3.1 A SCHEMATIC DIAGRAM INTRODUCING NOISE SOURCES IN THE STATIC MODEL.	73
FIG. 3.2 THE STATIC SIMULATION STEPS.	75
FIG. 3.3 THE CONCEPT OF CHARGE COLLECTION BY NEIGHBOURING PIXELS IMPLEMENTED IN SIMULATIONS.	76
Fig. 3.4 The map of counts for the scan of the $250~\mu$ m x $250~\mu$ m ROI obtained in simulations of a detector we	)RKING A) IN
THE SPC MODE B) IN THE C8P1 MODE FOR THE 8 KEV PHOTON ENERGY AND THE 4 KEV THRESHOLD SETTING	
Fig. 3.5 The maps of counts for the scan of the $250 \ \mu\text{m} \ x \ 250 \ \mu\text{m}$ ROI for the test variants listed in Table 3.1.	
FIG. 3.6 THE MEASUREMENT CONDITIONS WHEN THE PULSE INJECTION CIRCUITRY INJECTS TWO DIFFERENT CALIBRATION	PULSES INTO
NEIGHBOURING PIXELS P1 AND P2 EMULATING X-RAY BEAM MOVEMENT ALONG THE HORIZONTAL AXIS	
FIG. 3.7 THE NUMBER OF COUNTS VERSUS THRESHOLD AND DISTANCE FROM THE PIXEL BORDER IN A) THE SPC MODE B) THE (	C8P1 mode.
PIXEL P1 IS REPRESENTED WITH BLUE AND P2 WITH ORANGE COLOUR	
FIG. 3.8 THE NUMBER OF MISTAKES, NAMELY, THE EXTRA HITS COUNTED MARKED IN YELLOW, AND THE HITS LOST MARKED IN	BLUE IN THE
А) SPC моде, в) C8P1 моде	
FIG. 3.9 THE SIMPLIFIED READOUT CHANNEL ARCHITECTURE WITH THE C8P1 ALGORITHM FOR DYNAMIC SIMULATIONS	
FIG. 3.10 THE NOISE PEAKS REGISTERED IN DYNAMIC SIMULATIONS IN THE SPC AND C8P1 MODES	
FIG. 3.11 THE TEST SCENARIOS AND THRESHOLD SCANS REGISTERED IN THE DYNAMIC SIMULATIONS IN THE SPC AND C8P1	MODES FOR
PHOTON INTERACTIONS A) IN THE CENTRAL PIXEL AREA, B) AT THE BORDER BETWEEN TWO PIXELS	
FIG. 4.1 THE PHOTO OF THE CHASE JR. CHIP A) SINGLE, B) WITH THE BUMP-BONDED SILICON DETECTOR, C) THE LAYOUT OF TH	he Chase Jr.
CHIP. $1$ - PIXEL MATRIX, $2$ – DECOUPLING CAPACITORS AND ANALOG REFERENCES BLOCK, $3$ – LVDS TRANSCEIVERS, $4$ – $4$	ANALOG AND
DIGITAL PADS, 5 – DETECTOR GUARD RING, 6 – TESTS STRUCTURES	
FIG. 4.2 THE SINGLE READOUT CHANNEL ARCHITECTURE OF THE CHASE JR. CHIP WITH THE TESTABILITY FEATURES MARKED IN	N GREEN AND
THE TRIMMING FEATURES MARKED IN RED.	
FIG. 4.3 THE ANALOG CHAIN CONFIGURATION FOR THE PURPOSE OF THE DC OFFSETS TRIMMING	

Fig. 4.4 A) The trim DAC characteristics. B) The DC offset spread ( $\Sigma(V_T)$ ) as a function of the target DC offset. c) The
DC OFFSET SPREAD BEFORE AND AFTER TRIMMING
FIG. 4.5 THE ANALOG CHAIN CONFIGURATION FOR THE PURPOSE OF CSA GAINS TRIMMING
FIG. 4.6 THE CSA GAIN VALUES FOR ALL BCSA PARAMETER CONFIGURATIONS FOR SELECTED PIXELS
FIG. 4.7 THE ANALOG CHAIN CONFIGURATION FOR THE PURPOSE OF SH SLOW GAINS TRIMMING
FIG. 5.1 THE SETUP REQUIRED FOR THE ASIC TESTS
FIG. 5.2 THE USER INTERFACE PANEL ALLOWING SETTING THE CHIP CONFIGURATION.
FIG. 5.3 THE USER INTERFACE PANEL FOR THE THRESHOLD SCAN TESTS
FIG. 5.4 THE APPLICATION ARCHITECTURE DIAGRAM
FIG. 5.5 THE QUEUED STATE MACHINE CONCEPT IMPLEMENTED IN LABVIEW
FIG. 6.1 THRESHOLD SCANS FOR 432 PIXELS A) BEFORE AND B) AFTER DC OFFSETS CORRECTION. C) HISTOGRAMS OF THE DC OFFSETS
OF THE DISCRIMINATOR THRESHOLDS FOR $432$ pixels before and after correction
FIG. 6.2 THE CSA GAINS BEFORE AND AFTER CORRECTION. A) GAIN VALUES FOR ALL THE PIXELS, B) GAINS HISTOGRAM
FIG. 6.3 THE SLOW PATH GAINS BEFORE AND AFTER CORRECTION. A) GAIN VALUES FOR ALL THE PIXELS, B) GAINS HISTOGRAM
FIG. 6.4 THRESHOLD SCANS FOR THE DETECTOR ILLUMINATED WITH X-RAYS A) BEFORE AND B) AFTER CSA CORRECTION
FIG. 6.5 THE OUTPUT PULSE AMPLITUDE MEASURED FOR 432 PIXELS VERSUS IMPULSE CALIBRATION AMPLITUDE, A) RAW DATA, B) LINEAR
FITTING MODELS.
FIG. 6.6 A) THE MEASUREMENT CONDITIONS, WHERE PULSE INJECTION CIRCUITRY INJECTS FOUR DIFFERENT CALIBRATION PULSES INTO
NEIGHBOURING PIXELS P1, P2, P3, P4, REFLECTING CHARGE SHARING BETWEEN FOUR PIXELS. B) THRESHOLD SCANS IN THE SPC
MODE. C) THRESHOLD SCANS IN THE SUMMING AND THE C8P1 MODES
FIG. 6.7 FOUR DIFFERENT CALIBRATION PULSES ARE INJECTED IN THE CLUSTERS OF THE NEIGHBORING PIXELS. NUMBER OF COUNTS
MEASURED IN THE PIXEL MATRIX WHEN THE CHIP OPERATES IN A) THE SPC MODE, B) THE C8P1 MODE
FIG. 6.8 THE NUMBER OF COUNTS MEASURED IN THE PIXEL MATRIX WHEN THE CHIP OPERATES IN THE C8P1 MODE A) BEFORE
CORRECTION, B) AFTER DC OFFSETS CORRECTION, C) AFTER DC OFFSETS AND GAINS CORRECTION
FIG. 6.9 SCHEMATICALLY PRESENTED THRESHOLD SCANS FOR THREE PIXELS P1, P2, P3 IN THE CASE OF THE LARGE DC OFFSET SPREAD.
FIG. 6.10 THE MEASUREMENT CONDITIONS WHEN THE PULSE INJECTION CIRCUITRY INJECTS TWO DIFFERENT CALIBRATION PULSES INTO
THE NEIGHBOURING PIXELS P1 AND P2 EMULATING X-RAY BEAM MOVEMENT ALONG THE AXIS
FIG. 6.11 A) THE NUMBER OF COUNTS VERSUS THRESHOLD AND DISTANCE FROM THE PIXEL BORDER IN A) THE SPC MODE B) THE C8P1
MODE. THE PIXEL P1 IS REPRESENTED WITH BLUE AND P2 WITH ORANGE COLOUR
FIG. 6.12 THE EXPERIMENT SET-UP FOR THE TESTS WITH SYNCHROTRON RADIATION AT THE APS FACILITY
FIG. 6.13 THE NUMBER OF COUNTS MEASURED DURING THE EXPERIMENTS A) TEST 1, TEST 2, B) TEST 3, TEST 4 WITH A VISIBLE STEADY
COUNTS DECREASE DUE TO THE MONOCHROMATOR COOLING OFF
FIG. 6.14 THE PHOTOGRAPHS OF THE MEASUREMENT SET-UP USED FOR THE FLAT FIELD ILLUMINATION TESTS OF A DETECTOR WITH THE
Chase Jr. Chip

Fig. 6.15 The measurement assumptions: the photon interactions occur close to the border between neighbouring
PIXELS P1, P2. CHARGE IS SHARED BETWEEN THESE TWO PIXELS. B) THRESHOLD SCANS FOR THE PIXELS P1, P2 IN THE SPC MOI
(BLUE PLOTS) AND IN THE C8P1 MODE (RED PLOTS)12
Fig. 6.16 A) The measurement assumptions: the photon interactions occur close to the corner between the
NEIGHBOURING PIXELS P1, P2, P3, P4. CHARGE IS SHARED BETWEEN FOUR PIXELS. B) THRESHOLD SCANS FOR THE PIXELS P1, P
P3, P4 in the SPC mode (blue plots) and the C8P1 mode (red plots)12
FIG. 6.17 THE IDEA OF THE ROI SCAN WITH A PENCIL BEAM12
Fig. 6.18 The map of counts for the scan of the 700 µm x 700 µm ROI in the C8P1 mode A) before and B) after the
INTENSITY CORRECTION. THE CROSS-SECTIONS PARALLEL TO X AND Y AXES C) BEFORE AND D) AFTER THE INTENSITY CORRECTIO
129
Fig. 6.19 The map of counts for the scan of the 250 $\mu$ m x 250 $\mu$ m ROI in the C8P1 mode a) with LD = 0, b) LD = 1512
Fig. 6.20 The map of counts for the scan of the 250 $\mu$ m x 250 $\mu$ m ROI a) in the C8P1 mode b) in the SPC mode. c) ROI divisio
into two areas, namely, the pixel borders and the central pixel area. The cross-section parallel to X axis for Y=6
$\mu\text{M}$ and Y =85 $\mu\text{M}$ for d) the C8P1 mode and e) the SPC mode
Fig. $6.21$ The histograms comparing the numbers of counts in a) the central pixels areas for the ${\sf C8P1}$ and the SPC mod
B) THE CENTRAL PIXELS AREAS AND THE BORDER AREAS FOR THE SPC MODE, C) THE CENTRAL PIXELS AREAS AND THE BORDER ARE
FOR THE C8P1 MODE
Fig. 6.22. A) The map of counts for the scan of the $700~\mu$ m x $700~\mu$ m ROI in the C8P1 mode for uncorrected both CSA gain
AND SH SLOW GAINS AND B) A CORRESPONDING CROSS-SECTION. C) THE MAP OF COUNTS FOR CORRECTED CSA GAINS AN
UNCORRECTED SH SLOW GAINS AND D) A CORRESPONDING CROSS-SECTION.
FIG. 6.23. THE OUTPUT COUNT RATES PLOTTED AS A FUNCTION OF THE INPUT COUNT RATES FITTED TO THE PARALYSABLE MODEL FOR TH
CHIP OPERATING IN A) THE SPC MODE AND B) THE C8P1 MODE

### LIST OF TABLES

TABLE 1.1 THE APS AT ARGONNE NATIONAL LABORATORY [6] RADIATION CHARACTERISTICS	28
TABLE 1.2. THE PROPERTIES OF THE MOST POPULAR SENSOR MATERIALS AT 300 K.	42
TABLE 1.3. THE THRESHOLD VOLTAGE DISPERSION $\Sigma(V_T)$ for selected two technology nodes and two transistor sizes	47
TABLE 1.4. A SAMPLE DETECTOR PARAMETERS ASSUMED FOR ENC CALCULATIONS BASED ON [56], [66]	50
TABLE 2.1 THE ASICS WITH CHARGE RECONSTRUCTION ALGORITHMS, ACCORDING TO [10], [19], [22], [24], [34]	58
TABLE 3.1. INPUT PARAMETERS FOR THE STATIC DETECTOR MODEL	79
TABLE 4.1 THE LIST OF THE CHASE JR. CONFIGURATION BITS USED FOR TESTING AND TRIMMING PURPOSES.	94
TABLE 5.1 STANDARD TEST PROCEDURES FOR THE ASICS BEFORE BONDING TO THE SENSOR.	104
TABLE 5.2 A LIST OF MEASUREMENT SETUP DEVICES CHOSEN FOR THE CHASE JR. TESTS	105
TABLE 6.1 THE GAIN CORRECTION RESULTS.	113